

I. Formulation – Assumption – Proof – Practice

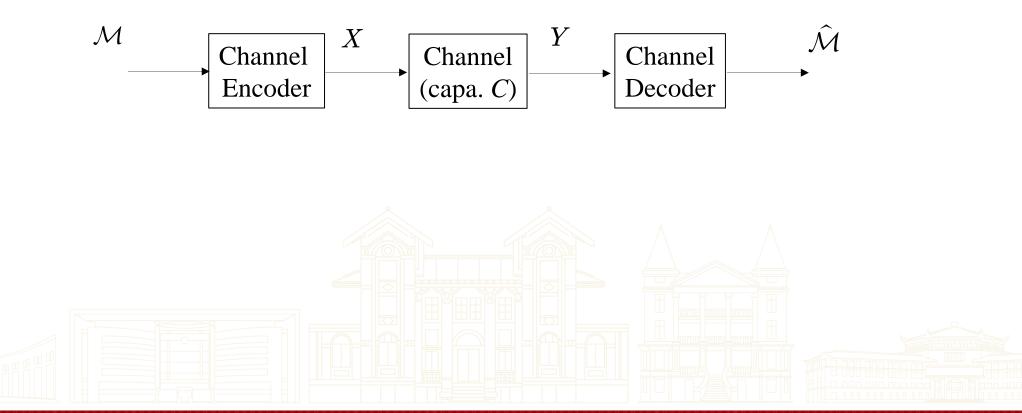
II. The Art of Channel Coding





I. Formulation

How to achieve channel capacity (C)?





I. Formulation

$$\begin{array}{c} \mathcal{M}(m^{k}) \\ \hline \\ \text{Channel} \\ \text{Encoder} \end{array} \begin{array}{c} X(x^{n}) \\ \hline \\ (\text{capa. } C) \end{array} \begin{array}{c} Y(y^{n}) \\ \hline \\ \text{Decoder} \end{array} \begin{array}{c} \hat{\mathcal{M}}(\hat{m}^{k}) \\ \hline \\ (y^{n} = x^{n} + e^{n}) \end{array}$$

Channel Coding: Adding redundancy into message can correct errors introduced by the channel $\dim(\mathcal{M}) < \dim(X)$

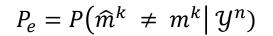
$$\operatorname{rate} = \frac{\dim(\mathcal{M})}{\dim(X)} = \frac{k}{n} = r$$

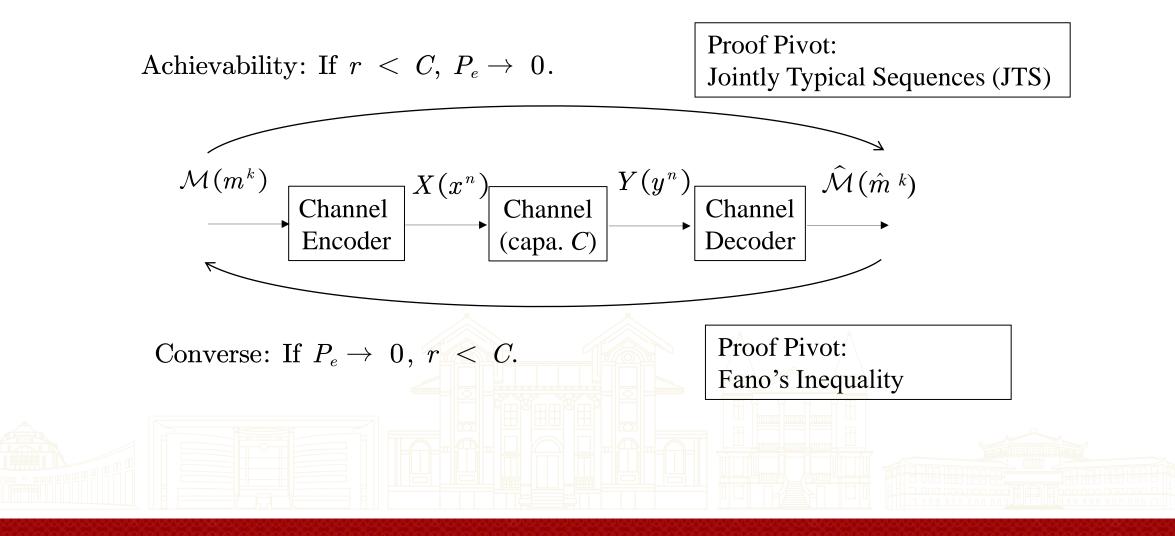
$$\underbrace{\begin{array}{c} & & \\ &$$



I. Formulation

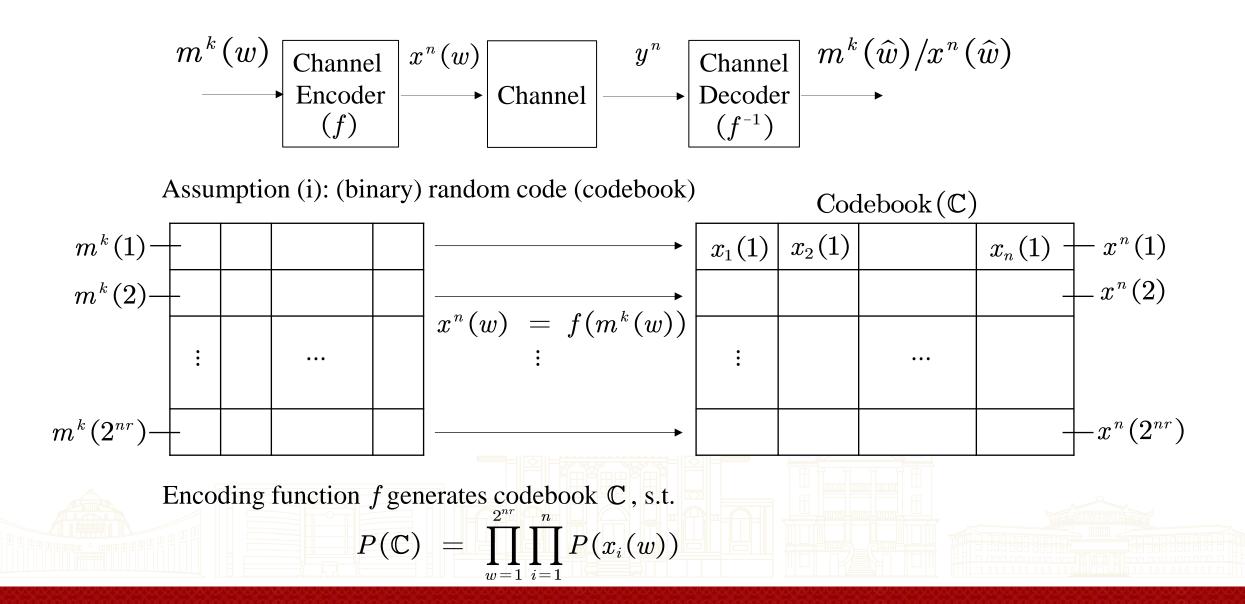
Shannon's Channel Coding Theorem







I. Formulation - Assumption





I. Formulation - Assumptions

$$\begin{array}{c|c} m^{k}(w) & \hline \text{Channel} \\ \hline & \text{Encoder} \\ (f) & \end{array} \begin{array}{c} x^{n}(w) & \hline & \text{Channel} \\ \hline & \text{Channel} \end{array} \begin{array}{c} y^{n} & \hline & \text{Channel} \\ \hline & \text{Decoder} \\ (f^{-1}) & \end{array} \begin{array}{c} m^{k}(\hat{w}) / x^{n}(\hat{w}) \\ \hline & & (f^{-1}) \end{array} \end{array}$$

Assumption (ii): Both sides know the channel

Assumption (iii): $m^{k}(w)(x^{n}(w))$ are uniformly chosen for transmission

$$P(x^n(w)) \;=\; P(m^k(w)) \;=\; rac{1}{2^{nr}}$$

Assumption (iv): The channel is discrete memoryless

$$P(y^{n}|x^{n}(w)) = \prod_{i=1}^{n} P(y_{i}|x_{i}(w))$$



I. Assumptions - Proof

Achievability: If $r \ < \ C, \ P_e
ightarrow \ 0.$

$$\xrightarrow{y^{n}} \begin{array}{c} \text{Channel} \\ \text{Decoder} \\ (f^{-1}) \end{array} \xrightarrow{m^{k}(\hat{w})/x^{n}(\hat{w})}$$

$$y^n \& x^n(\widehat{w})$$
 are a pair of JTS

JTS: Given
$$\epsilon \to 0$$
, y^n and $x^n(\hat{w})$ are JTS if
 $\left|-\frac{1}{n}\log_2 P(x^n(\hat{w})) - H(X)\right| < \epsilon$
 $\left|-\frac{1}{n}\log_2 P(y^n) - H(Y)\right| < \epsilon$
 $\left|-\frac{1}{n}\log_2 P(x^n(\hat{w}), y^n) - H(X,Y)\right| < \epsilon$



I. Assumptions - Proof

Achievability: If
$$r < C, P_e \rightarrow 0$$
.

$$\begin{array}{c|c} m^{k}(w) & \hline \text{Channel} \\ \hline & & \text{Encoder} \\ (f) & & & \end{array} \end{array} \xrightarrow{x^{n}(w)} \hline \text{Channel} \\ \hline & & & \text{Channel} \\ \hline & & & & \\ \end{array} \xrightarrow{y^{n}} \hline \text{Channel} \\ \hline & & & & \\ \hline & & & & \\ \end{array} \xrightarrow{y^{n}} \hline \text{Channel} \\ \hline & & & & \\ \hline & & & & \\ \hline & & & & \\ \end{array} \xrightarrow{y^{n}} \hline \text{Channel} \\ \hline & & & & \\ \hline & & & & \\ \hline & & & & \\ \end{array} \xrightarrow{y^{n}} \hline \begin{array}{c} \text{Channel} \\ \text{Decoder} \\ (f^{-1}) \end{array} \xrightarrow{y^{n}(\hat{w})} x^{n}(\hat{w})$$

JTS property (1): If $x^n(w)$ and y^n are drawn i.i.d, s.t. $P(x^n(w), y^n) = \prod_{i=1}^n P(x_i(w), y_i)$, ensured by Assumptions (i) (iv) When $n \to \infty$, $P(x^n(w) \text{ and } y^n \text{ are JTS}) = 1 - \epsilon$, JTS property (2): If $x^n(w)$ and y^n are independent, i.e., $P(x^n(w), y^n) = P(x^n(w)) \cdot P(y^n)$, $P(x^n(w) \text{ and } y^n \text{ are JTS}) \leq 2^{-n(I(X;Y)-3\epsilon)}$



I. Assumptions - Proof

Achievability: If $r < C, P_e \rightarrow 0$.

$$P_{e} = \sum_{\mathbb{C}} P(\mathbb{C}) P_{e}(\mathbb{C})$$

$$= \frac{1}{2^{nr}} \sum_{\mathbb{C}} P(\mathbb{C}) \cdot \sum_{w=1}^{2^{nr}} P_{e,w}(\mathbb{C})$$

$$\text{code construction symmetry} = \sum_{\mathbb{C}} P(\mathbb{C}) \cdot P_{e,1}(\mathbb{C})$$

$$= P_{e,1}$$

$$P_{e}(\mathbb{C}) - \text{Error probability of code } \mathbb{C}$$

$$P_{e,w}(\mathbb{C}) - \text{Error probability of code word } x^{n}(w)$$



I. Assumptions - Proof

 $ext{Achievability: If } r \ < \ C, \ P_e
ightarrow \ 0.$

$$E_{w}^{:} x^{n}(w) \text{ and } y^{n} \text{ are JTS.}$$

$$P_{e,1} = P(E_{1}^{C} \cup E_{2} \cup \dots \cup E_{2^{nr}})$$

$$\leq \epsilon + \sum_{w=2}^{2^{nr}} 2^{-n(I(X;Y)-3\epsilon)} \quad Properties (1)(2) \text{ of JTS}$$

$$= \epsilon + (2^{nr} - 1) \cdot 2^{-n(I(X;Y)-3\epsilon)}$$

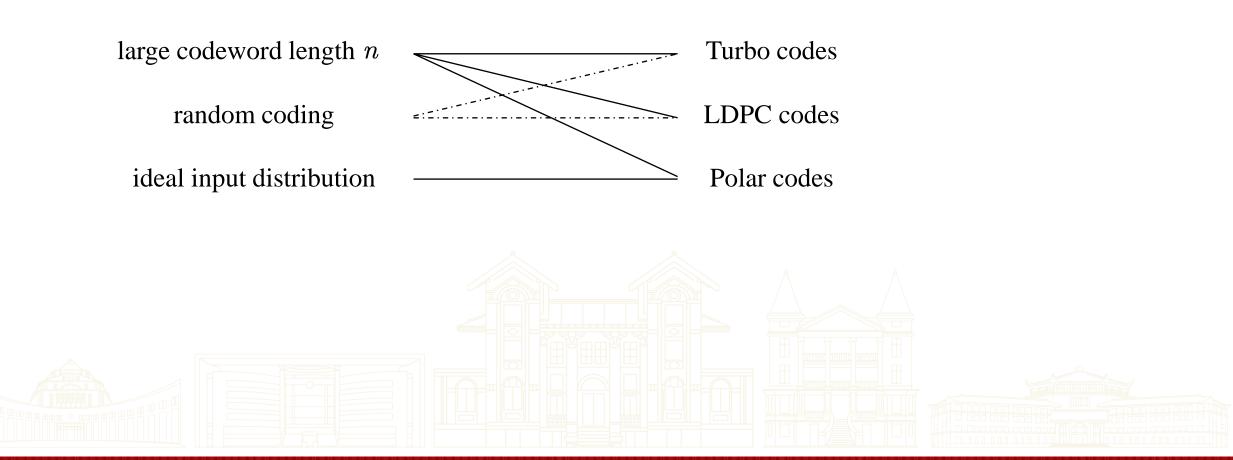
$$< \epsilon + 2^{-n(I(X;Y)-r-3\epsilon)}$$
IF input distribution is ideal, s.t. $I(X;Y) = C$ Assumption (i)
$$IF r < C \& n \to \infty$$

$$P_{e} = P_{e,1} = 2\epsilon$$



I. Proof - Application

Disparity between Assumptions and Practice.





I. Assumptions - Proof

Converse: If $P_e \rightarrow 0, r < C$.

 $I(m^k(w); y^n) = H(m^k(w)) - H(m^k(w)|y^n)$ $H(m^k(w)) = \log_2 2^{nr} = nr$ Assumption (iii) $I(m^k(w); y^n) \leq I(x^n(w); y^n)$ Data Processing Inequality $\leq n \cdot C$ Assumptions (i) (iv) $H(m^{k}(w)|y^{n}) = H(c^{n}(w)|y^{n})$ $\leq H(P_e) + P_e \log_2(2^{nr} - 1)$ Fano's Inequality $\lesssim 1 + P_e \cdot nr$ $nr \leq 1 + P_e \cdot nr + nC$, or $r \leq C + \frac{1}{n} + P_e \cdot r$ IF $P_e \to 0 \& n \to \infty$ $r \leq C$



II. The Art of Channel Coding

$$\begin{array}{c|c} m^{k}(w) & \hline \text{Channel} \\ \hline & \text{Encoder} \\ (f) & \end{array} \end{array} \xrightarrow{x^{n}(w)} \hline \text{Channel} & y^{n} & \hline \text{Channel} \\ \hline & \text{Decoder} \\ (f^{-1}) & \end{array} \xrightarrow{m^{k}(\hat{w})}$$

Random coding is Good for proof, but Bad for Practice

 $egin{array}{lll} f\colon m^{\,k}(w) &\mapsto x^n(w) \ f^{-1}\colon y^n &\mapsto m^k(\widehat{w})/\!\!x^n(\widehat{w}) \end{array}$

The art of channel coding arises as we refrain f from randomness



II. The Art of Channel Coding

In light of linear block codes,

$$egin{aligned} f:\mathbf{G}&=egin{bmatrix} x^n\left(1
ight)\ x^n\left(2
ight)\ dots\ x^n\left(k
ight)\end{bmatrix}\ x^n\left(k
ight)\ x^n\left(k
ight)\end{bmatrix}\ x^n\left(w
ight)&=m^k\left(w
ight)\cdot\mathbf{G}.\ &\mathbb{C}&=&\{m^k\left(w
ight)\cdot\mathbf{G}\}. \end{aligned}$$

a basis of k linearly independent codewords

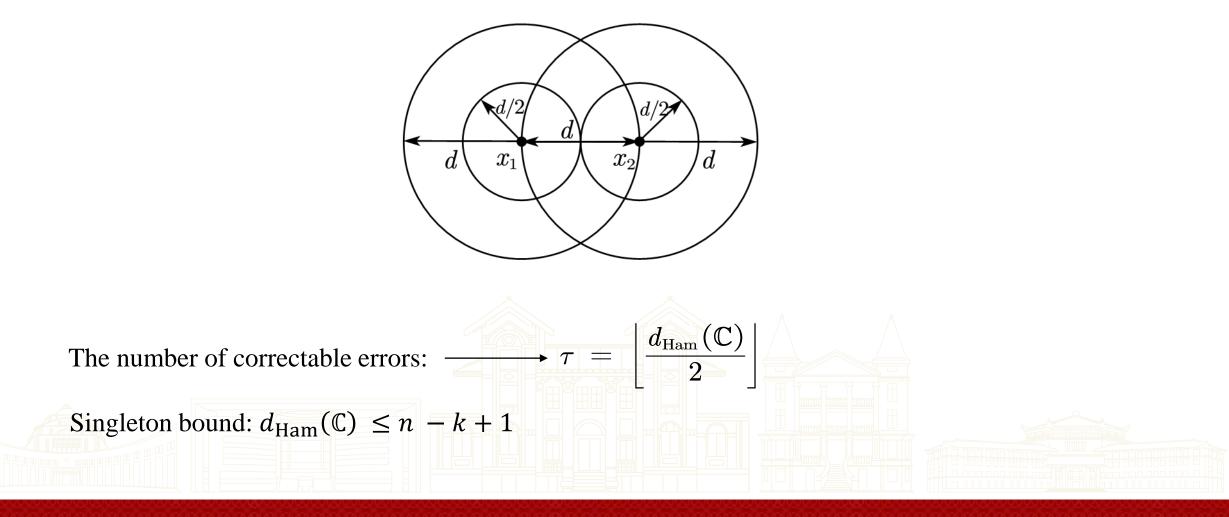
$$egin{array}{rll} x^n(w)&=&m^k(w)\cdot\,\,{f G}.\ &{\Bbb C}&=&\{m^k(w)\,\cdot\,\,{f G},\,\,orall w\} \end{array}$$

If **G** is a semi-definite and tridiagonal matrix, the code becomes a convolutional code



II. The Art of Channel Coding

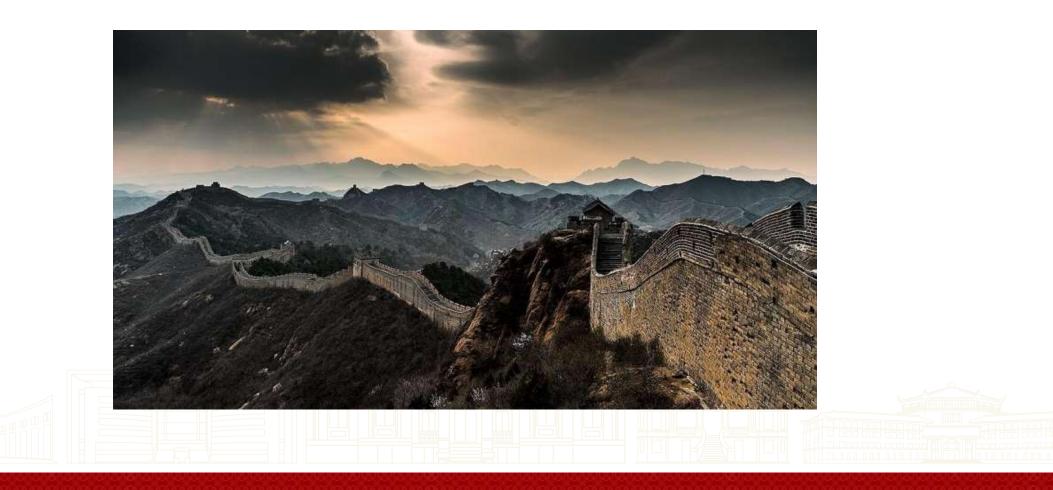
For an (n, k) block code $(\mathbb{C}), d_{\operatorname{Ham}}(\mathbb{C}) = \min\{d_{\operatorname{Ham}}(x^n(w), x^n(w'))\}$





- II. The Art of Channel Coding
 - (n, 1, n) Repetition code $\mathbf{G} = (1, 1, \dots, 1)$

Majority voting realizes maximum likelihood (ML) decoding

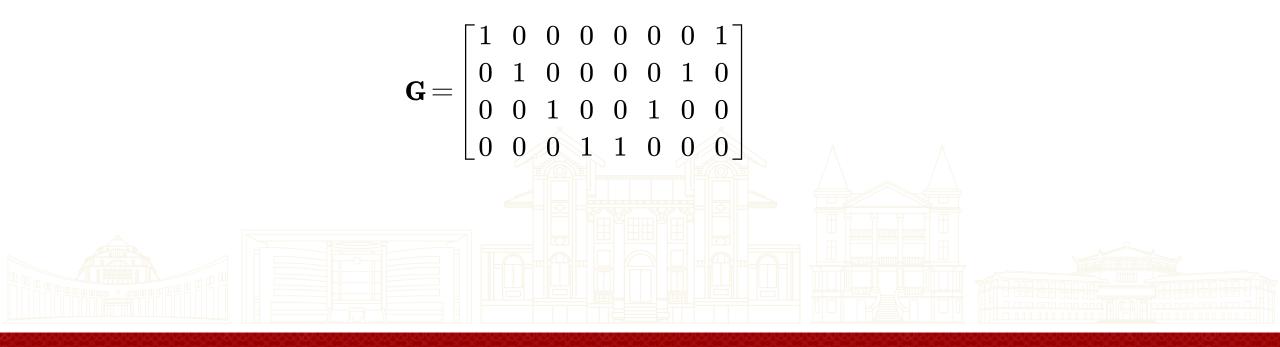




II. The Art of Channel Coding

Symmetric Code for Poetry

(中秋圆月) · G = (中秋圆月 月圆秋中)
(围席似月) · G = (围席似月 月似席围)





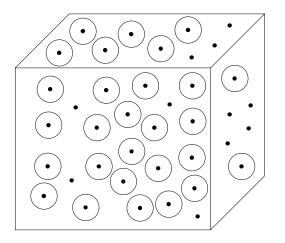
II. The Art of Channel Coding

$$(n, k)$$
 block code $\perp (n, n-k)$ block code

(dual codes)

$$f_{\scriptscriptstyle \perp}\,:\, {f G}_{\scriptscriptstyle \perp} \ = \ egin{bmatrix} x_{\scriptscriptstyle \perp}^{\,n}\,(1)\ x_{\scriptscriptstyle \perp}^{\,n}\,(2)\ dots\ x_{\scriptscriptstyle \perp}^{\,n}\,(n-k) \end{bmatrix}$$

(a basis of (n - k) linearly independent dual codewords)



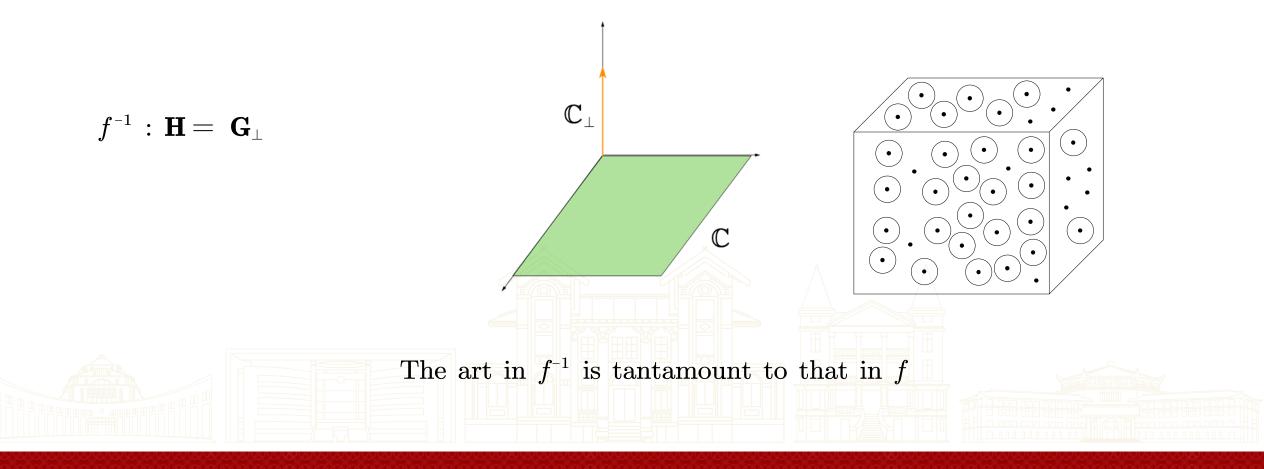
$$x_{\perp}^{n}(w) = m^{n-k}(w) \cdot \mathbf{G}_{\perp}, \ x^{n}(w) \cdot (x_{\perp}^{n}(w))^{\mathrm{T}} = 0$$
 $\mathbb{C}_{\perp} = \{m^{n-k}(w) \cdot \mathbf{G}_{\perp}, \forall w\}$



II. The Art of Channel Coding

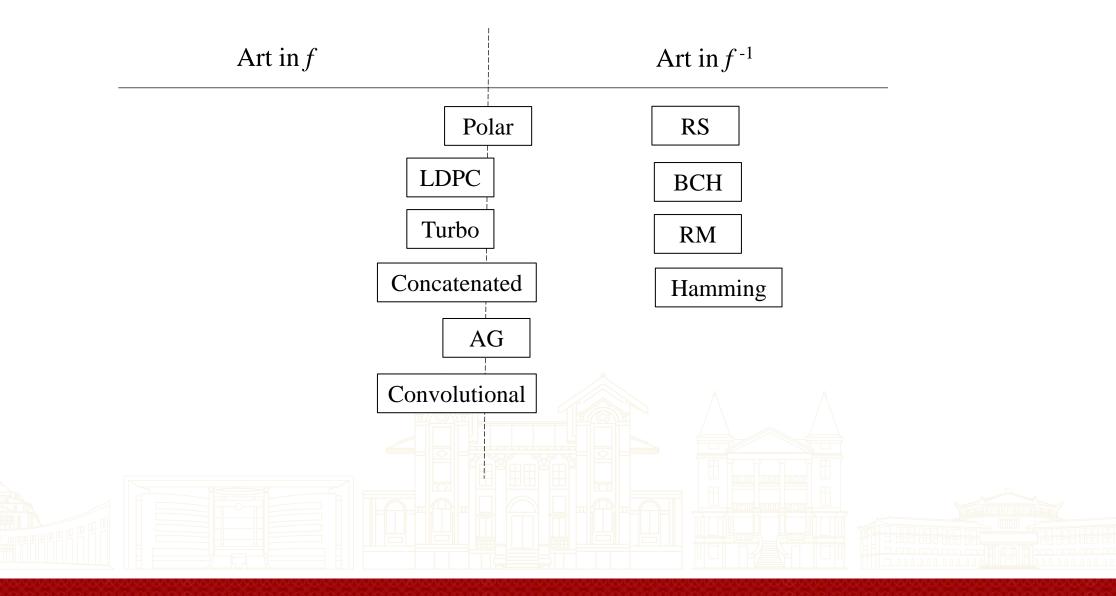
(n, k) block code $\perp (n, n-k)$ block code

(dual codes) \mathbb{C} and \mathbb{C}_{\perp} are orthogonal





II. The Art of Channel Coding





References:



- 1. Website: <u>www.chencode.cn</u>
- 2. Email: <u>chenli55@mail.sysu.edu.cn</u>
- 3. Videos: <u>https://www.xuetangx.com/course/sysu0807bt</u>

信息论与编码

